

# Cricket: A Self-Powered Chirping Pixel Supplemental Material

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## ACM Reference Format:

Shree K. Nayar, Jeremy Klotz, Nikhil Nanda, and Mikhail Fridberg. 2024. Cricket: A Self-Powered Chirping Pixel Supplemental Material. *ACM Trans. Graph.* 43, 4, Article 151 (July 2024), 2 pages. <https://doi.org/10.1145/3658196>

## 1 CRICKET CIRCUIT DETAILS

Here we provide a detailed explanation of the cricket circuit in Fig. 1. When the photovoltaic cell is exposed to light, it charges capacitor  $C_1$ , and voltage  $V_c$  begins to rise. At some voltage, the comparator comes to life. At that point, the reference (regulated) voltage  $V_r$  is smaller than the voltage  $V_1$  and hence the output  $V_o$  of the comparator is 0. Since  $V_c$  continues to rise, at some point,  $V_1$  will exceed  $V_2$ , and the comparator output  $V_o$  goes from 0 to  $V_c$ . This activates the oscillator, which puts out an RF frequency (in the GHz range). The frequency  $f_{id}$  of the oscillator output, which is the identifier of the cricket, is determined by the voltage  $V_f$ , which, in turn, is determined by the resistors  $R_t$  and  $R_b$ , and the capacitor  $C_2$ . In our implementation,  $f_{id}$  is preset by selecting the resistor  $R_b$ .

Although the oscillator is now active, it is not yet connected to the antenna due to the switch  $S$ . The closing of this switch is delayed by  $R_4$  and  $C_3$ . This delay is introduced to allow the preset voltage  $V_f$  to stabilize and ensure the frequency applied to the antenna is precise and stable. In the current implementation, this delay has been set to roughly 10  $\mu$ s. After the delay, the oscillator is connected to the antenna, which transmits an RF chirp for about 30  $\mu$ s. This chirp duration is limited solely by the fact that the oscillator is, by far, the highest power consumer in the circuit. Hence, while the oscillator is active,  $V_c$  falls rapidly until  $V_1$  goes lower than  $V_2$ , the comparator output goes to 0, and the oscillator shuts down. At this point,  $C_1$  begins to recharge and  $V_c$  rises again.

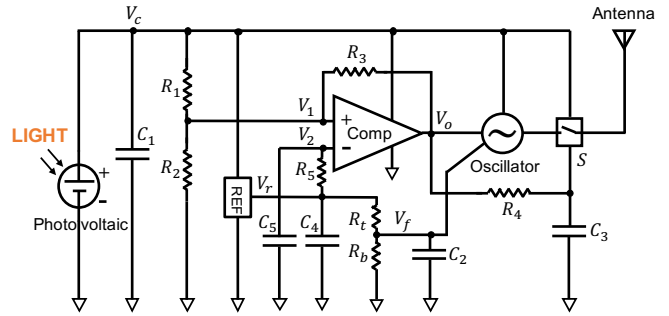


Fig. 1. Cricket circuit.

Table 1. List of components in the cricket circuit.

Component	Value
$C_1$	3.3 $\mu$ F
$C_2$	10 $\mu$ F
$C_3$	220 pF
$C_4$	0.1 $\mu$ F
$C_5$	1 $\mu$ F
$R_1$	6.2 M $\Omega$
$R_2$	10 M $\Omega$
$R_3$	100 M $\Omega$
$R_4$	100 k $\Omega$
$R_5$	1 M $\Omega$
$R_t$	4.02 M $\Omega$
$R_b$	10 – 100 M $\Omega$ (adjustable)
Photovoltaic	Panasonic AM-5610CAw-DGK-T
Voltage Reference	ABLIC Inc. S-1318D18-M5T1U4
Comparator	Texas Instruments TLV3691
Oscillator	Analog Devices MAX2752
Switch	Analog Devices ADG 901
Antenna	Taoglas FXP 29 (2.04 – 2.10 GHz)

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 0730-0301/2024/7-ART151 \$15.00  
<https://doi.org/10.1145/3658196>

Our current implementation uses off-the-shelf passive (resistors and capacitors) and active (comparator, oscillator and switch) components. The values and model numbers of the components we used in our prototype are given in Tab. 1. The resistor  $R_3$  introduces hysteresis in the comparator output. It can be lowered to increase the length of the chirp. In our case, however, we want the chirp to be very short and hence we have used a large value (100 M $\Omega$ ).

Table 2. Current drain of components in the cricket circuit.

Component	Chirping (nA)	Sleeping (nA)
Oscillator	$10^7$	200
Voltage divider, $R_1$ and $R_2$	181	181
Voltage divider, $R_t$ and $R_b$	116	116
Switch	100	100
Voltage reference	95	95
Comparator	75	75
Total	$10^7$	767

## 2 POWER MODEL DETAILS

We have estimated the power consumption of a cricket from the typical current drain of each of its components. In Tab. 2, we list the current drain for each component when the cricket when the cricket is chirping and sleeping (not chirping). When the cricket is chirping, the oscillator consumes 10 mA during its 40  $\mu$ s chirp. When the cricket is sleeping, the active components and voltage dividers consume 767 nA.

The power consumed by each component is also function of the voltage. We will assume a constant voltage  $V_c = 2.9$  V to compute the power consumption from the currents in Tab. 2.<sup>1</sup> This is a valid choice for all components, including those powered by  $V_r$ , because the voltage reference is a linear regulator. Using this voltage, the power consumed while chirping and sleeping is:

$$P_{chirping} = 29.0 \text{ mW}, \quad P_{sleeping} = 2.23 \text{ } \mu\text{W}. \quad (1)$$

A typical cricket chirps roughly 0.1 chirps per lux, per second, and each chirp lasts 40  $\mu$ s. From these facts, we can compute the time spent chirping and sleeping as a function of the light level,  $L$  lux, in a time interval  $T$  seconds as:

$$T_{chirping} = 4LT \cdot 10^{-6} \text{ s}, \quad T_{sleeping} = T - T_{chirping} \text{ s}. \quad (2)$$

Now, the average power consumption is given by:

$$P = \frac{T_{chirping} \cdot P_{chirping} + T_{sleeping} \cdot P_{sleeping}}{T} \text{ W}. \quad (3)$$

We can rewrite  $P$  as a function of the light level,  $L$  lux, as  $P = \alpha L + \beta$ , where  $\alpha = 116$  nW/lux and  $\beta = 2.23$   $\mu$ W.

## 3 ETRANSITION GLASS CIRCUIT DETAILS

Here we detail the eTransition glass circuit in Fig. 2, a self-powered circuit inspired by cricket that drives a liquid crystal (LC) light valve. As mentioned in the main paper, the first part of the cricket circuit in Fig. 1 – from the photovoltaic to the comparator output – is essentially a self-powered pulse generator with a very low duty cycle. In place of the oscillator, we use an inverter that converts  $V_o$  to an AC pulse train in order to drive the LC valve. The transmittance of the LC valve is therefore inversely proportional to the incident light.

<sup>1</sup>Technically,  $V_c$  is not a constant but a sawtooth wave shown in Fig. 4 of the main paper. To approximate the power consumption, we conservatively use  $V_c = 2.9$  V, the threshold voltage at which the cricket chirps.

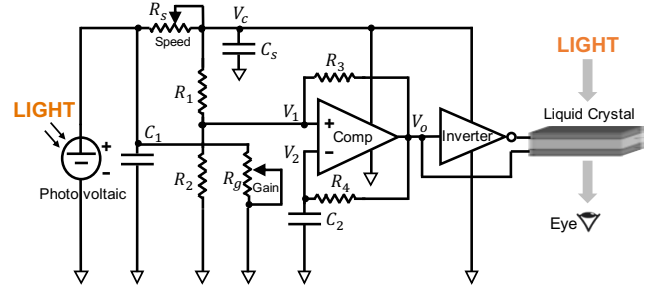


Fig. 2. eTransition glass circuit.

Table 3. List of components in the eTransition glass circuit.

Component	Value
$C_1$	3.3 $\mu$ F
$C_2$	100 pF
$C_s$	0.1 uF
$R_1$	10 M $\Omega$
$R_2$	10 M $\Omega$
$R_3$	10 M $\Omega$
$R_4$	10 M $\Omega$
$R_g$	100 M $\Omega$
$R_s$	1 k $\Omega$
Photovoltaic ( $\times 6$ )	Vishay BPW34
Comparator	Texas Instruments TLV3691
Inverter	Texas Instruments SN74AUP1G04
LC Light Valve ( $\times 2$ )	Adafruit PID 3627

As with the cricket, when the photovoltaic is exposed to light, the voltage  $V_c$  begins to rise. Initially, the comparator output  $V_o$  is 0. When  $V_1$  exceeds  $V_2$ ,  $V_o$  jumps to  $V_c$ . At this point,  $V_2$  begins to rise due to  $R_4$  and  $C_2$ . At some point,  $V_2$  exceeds  $V_1$ , and the comparator output  $V_o$  drops to 0. The comparator output  $V_o$  is therefore a square wave and its duty cycle is determined by  $R_1$ ,  $R_2$  and  $R_3$ . We have chosen these resistors such that the duty cycle is 50%. The frequency of the square wave is determined by  $R_4$  and  $C_2$ . Most importantly, the peak value of the square wave is proportional to the intensity of light falling on the photovoltaic. Table 3 lists the component values used in this circuit.

We now have  $V_o$  varying between 0 and  $V_c$ . However, we need an AC voltage which includes a reverse in polarity. This is achieved by using the inverter. Both the output of the comparator and the output of the inverter have a push-pull configuration. These outputs are used to, in effect, create an H-bridge, which results in the LC valve seeing an AC voltage that flips between  $(V_o, 0)$  and  $(0, V_o)$ , where  $V_o$  is proportional to the incident light intensity.